



# Complex Impedance

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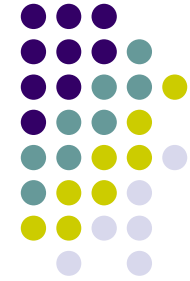


- ∅ The **Impedance Z** of a circuit is the ratio of phasor voltage **V** to the phasor current **I** (**Ohm's Law**) .which is usually represented by complex number.

$$Z = \frac{V}{I} \quad \text{or} \quad V = ZI \quad (\text{Ohm's Law})$$

- ∅ This **Impedance Z** is usually a complex number and not a sine wave.

# Complex Impedance in case of Resistance $Z_R$



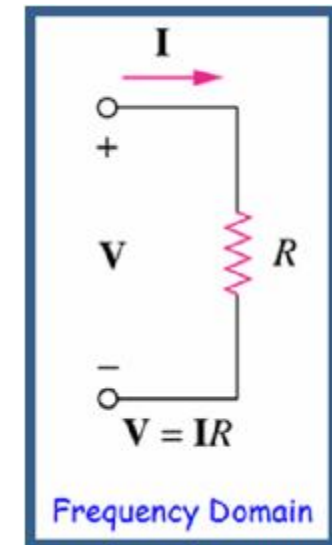
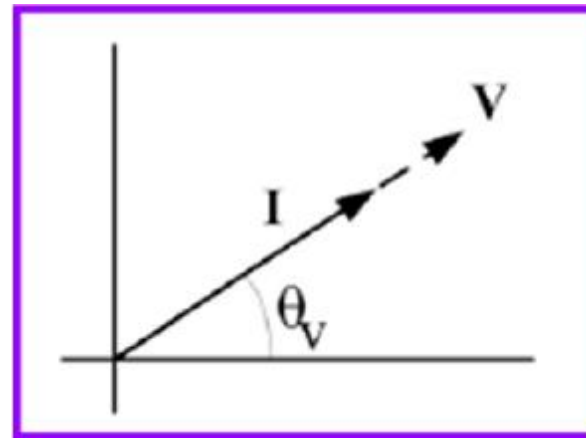
- ∅ In the Resistance circuit, the phasor current  $I$  is in phase with phasor voltage  $V$ .

$$V = V \angle \theta_v \quad I = I \angle \theta_v \quad (\text{current in phase with voltage})$$

Applying Ohm's law:

$$Z_R = \frac{V \angle \theta_v}{I \angle \theta_v} = \frac{V}{I} \angle 0^\circ$$

$$Z_R = R \angle 0^\circ = R$$



# Complex Impedance in case of Inductive reactance $Z_L$



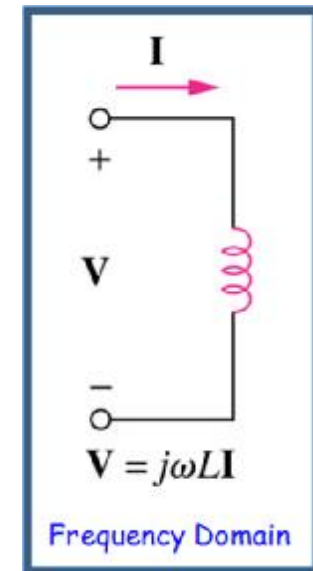
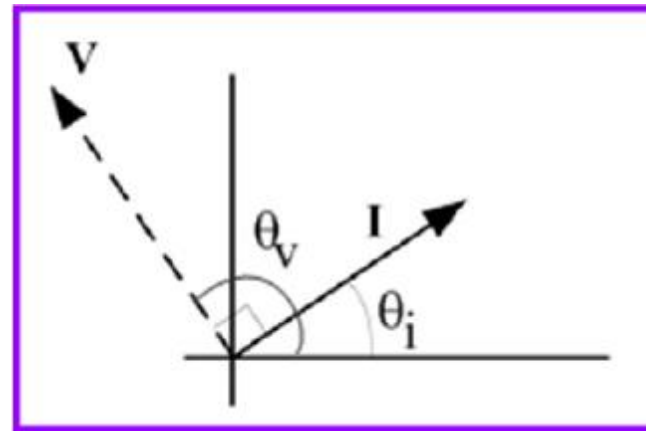
∅ In the Inductance circuit, the phasor current **I** is **LAGs** voltage **V** by **90°**.

$$V = V \angle \theta_v \quad I = I \angle (\theta_v - 90) \text{ (current lags voltage by } 90^\circ \text{)}$$

Applying Ohm's law:

$$Z_L = \frac{V \angle \theta_v}{I \angle (\theta_v - 90)}$$
$$= \frac{V}{I} \angle 90^\circ$$

$$Z_L = X_L \angle 90^\circ = jX_L$$



# Complex Impedance in case of Capacitive reactance $Z_C$

- ∅ In the Capacitive circuit, the phasor current **I** is **LEADS** voltage **V** by **90°**.

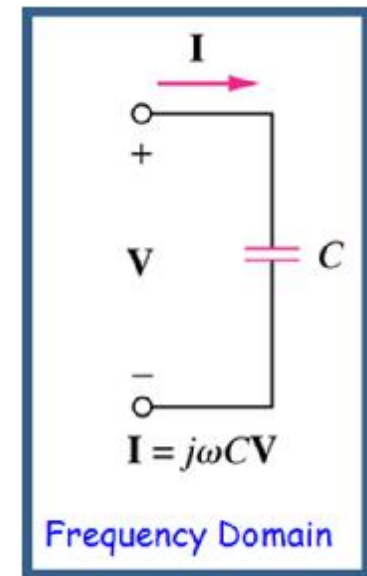
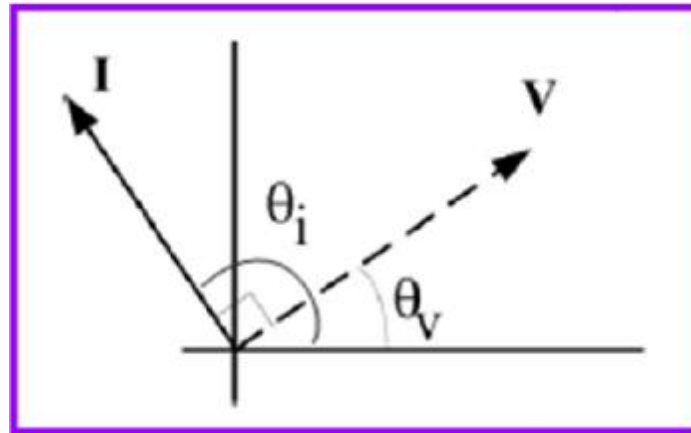
$$V = V \angle \theta_V \quad I = I \angle (\theta_V + 90^\circ) \quad (\text{current leads voltage by } 90^\circ)$$

Applying Ohm's law:

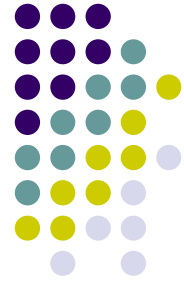
$$Z_C = \frac{V \angle \theta_V}{I \angle (\theta_V + 90^\circ)}$$

$$= \frac{V}{I} \angle -90^\circ$$

$$Z_C = X_C \angle -90^\circ = -jX_C$$



# Impedance of Joint Elements

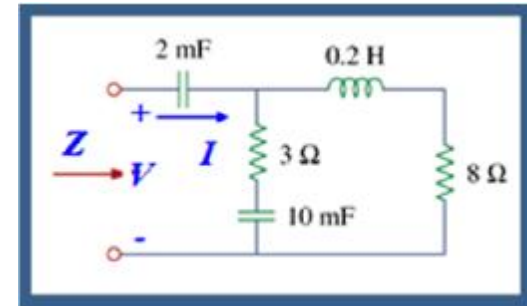


- ∅ The **Impedance Z** represents the opposition of the circuit to the flow of sinusoidal current.

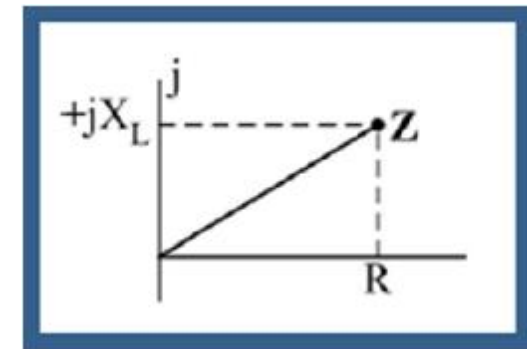
$$Z = \frac{V}{I} = R + jX =$$

=Resistance + j×Reactance

$$= |Z| \angle q$$



$$|Z| = \sqrt{R^2 + X^2} \quad q = \tan^{-1} \frac{X}{R}$$
$$R = |Z| \cos q \quad X = |Z| \sin q$$



- ∅ The Reactance is **Inductive** if **X is positive** and it is **Capacitive** if **X is negative**.

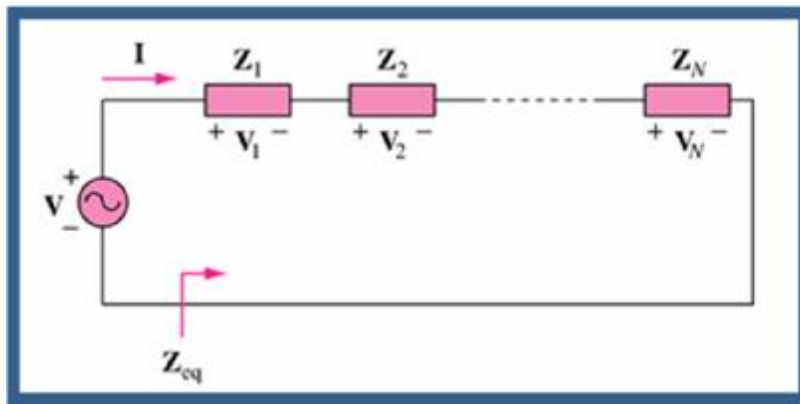


# Series and Parallel Circuits

# Series Circuits



- ∅ The Kirchoff's Voltage Law (KVL) holds in the frequency domain. For series connected impedances:



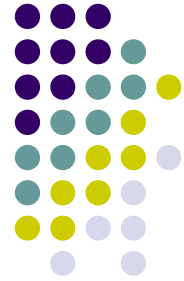
**Current is Constant**

$$V = V_1 + V_2 + \dots + V_N = I(Z_1 + Z_2 + \dots + Z_N)$$

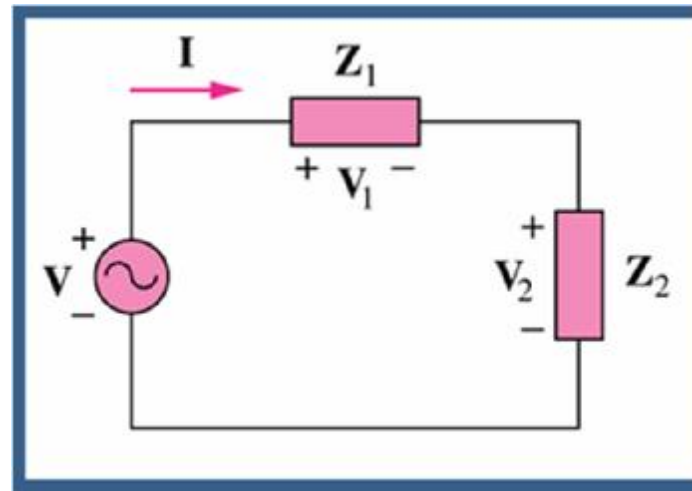
$$Z_{eq} = \frac{V}{I} = Z_1 + Z_2 + \dots + Z_N \quad (\text{Equivalent Impedance})$$



# Voltage Divider rule



∅ The **Voltage Division** for two elements in series is:



$$\mathbf{I} = \frac{\mathbf{V}}{\mathbf{Z}_1 + \mathbf{Z}_2}$$

Since  $\mathbf{V}_1 = \mathbf{Z}_1\mathbf{I}$  and  $\mathbf{V}_2 = \mathbf{Z}_2\mathbf{I}$ , then

$$\mathbf{V}_1 = \frac{\mathbf{Z}_1}{\mathbf{Z}_1 + \mathbf{Z}_2} \mathbf{V}$$
$$\mathbf{V}_2 = \frac{\mathbf{Z}_2}{\mathbf{Z}_1 + \mathbf{Z}_2} \mathbf{V}$$

# Admittance of Joint Elements



- ∅ The **Admittance Y** represents the admittance of the circuit to the flow of sinusoidal current. The admittance is measured in **Siemens (s)**

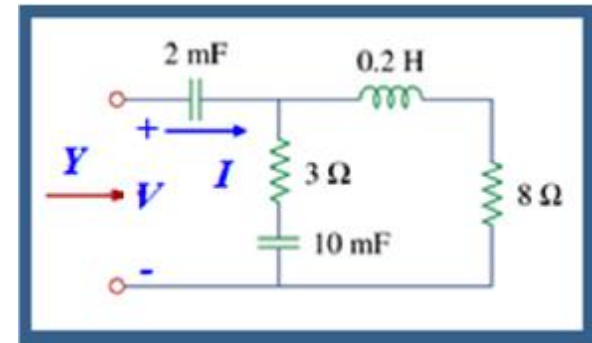
$$Y = \frac{1}{Z} = \frac{I}{V} = G + jB$$

= Conductance + j Suseptance =  $|Y| \angle \theta$

$$Y = G + jB = \frac{1}{R + jX} \frac{R - jX}{R - jX} = \frac{R - jX}{R^2 + X^2}$$

$$G = \frac{R}{R^2 + X^2}$$

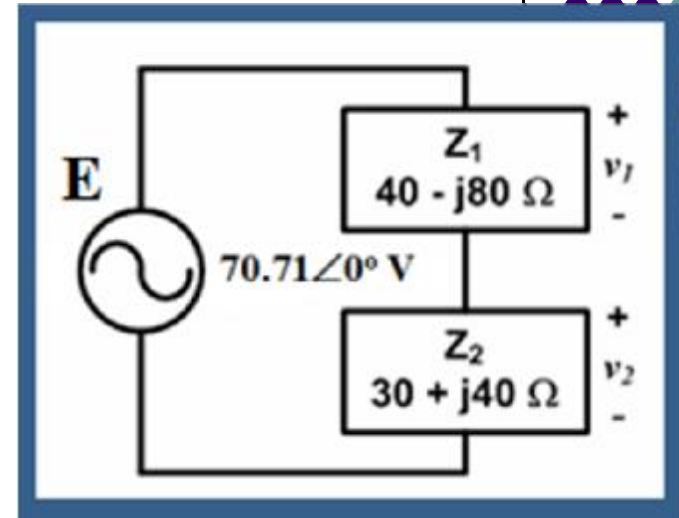
$$B = -\frac{X}{R^2 + X^2}$$



**Ex.5:** consider the circuit,

- (a) Calculate the sinusoidal voltages  $v_1$  and  $v_2$  using phasors and voltage divider rule  
 (b) Sketch the phasor diagram showing  $E$ ,  $V_1$  and  $V_2$ .

**Solution**



(a) the sinusoidal voltages  $v_1$  and  $v_2$  using phasors

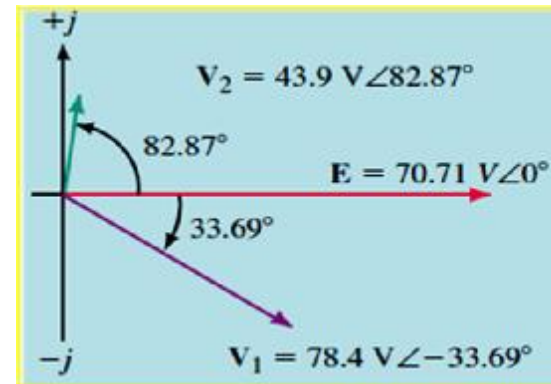
$$V_1 = \left( \frac{40 - j80}{(40 - j80) + (30 + j40)} \right) (70.71 \angle 0^\circ) = 78.4 \angle -33.69^\circ \text{ V}$$

$$V_2 = \left( \frac{30 + j40}{(40 - j80) + (30 + j40)} \right) (70.71 \angle 0^\circ) = 43.9 \angle 82.87^\circ \text{ V}$$

$$V_1 = \frac{Z_1}{Z_1 + Z_2} V$$

$$V_2 = \frac{Z_2}{Z_1 + Z_2} V$$

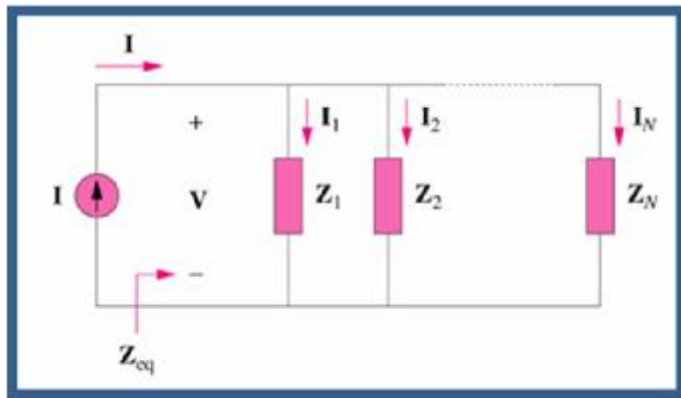
(b) Phasor diagram



# Parallel Circuits



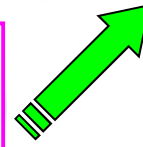
- ∅ The Kirchoff's Current Law (KCL) holds in the frequency domain. For Parallel connected impedances:



**Voltage is Constant**

(Equiv. Admitance)

$$I = I_1 + I_2 + \dots + I_N = V \left( \frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_N} \right)$$



$$\frac{1}{Z_{eq}} = \frac{I}{V} = \frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_N} \qquad Y_{eq} = Y_1 + Y_2 + \dots + Y_N$$

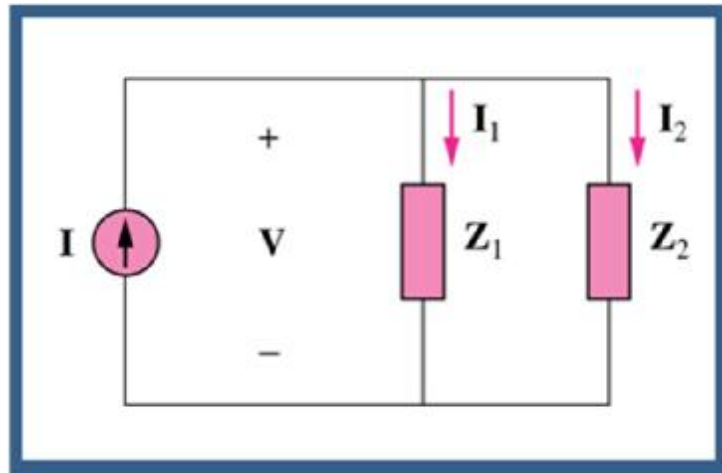
In case of two resistors

$$Z_{eq} = \frac{1}{Y_{eq}} = \frac{1}{Y_1 + Y_2} = \frac{1}{1/Z_1 + 1/Z_2} = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

# Current Divider rule



∅ The **Current Division** for two elements is:



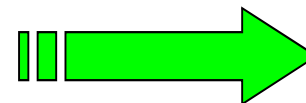
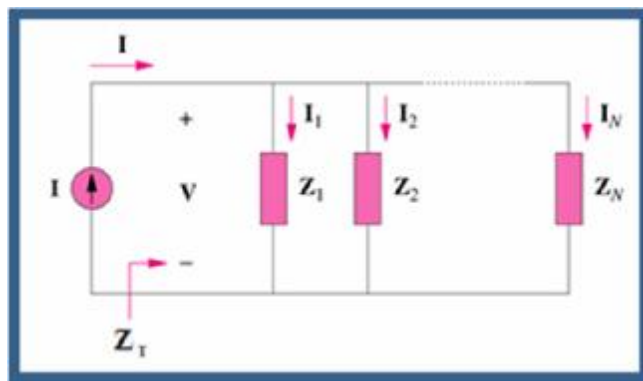
$$Z_{eq} = \frac{1}{Y_{eq}} = \frac{1}{Y_1 + Y_2} = \frac{1}{1/Z_1 + 1/Z_2} = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

$$V = I Z_{eq} = I_1 Z_1 = I_2 Z_2$$

$$I_1 = \frac{Z_2}{Z_1 + Z_2} I$$

$$I_2 = \frac{Z_1}{Z_1 + Z_2} I$$

∅ The **Current Division** for more elements is:



$$I_{Z_N} = \frac{Z_T}{Z_N} I_T$$

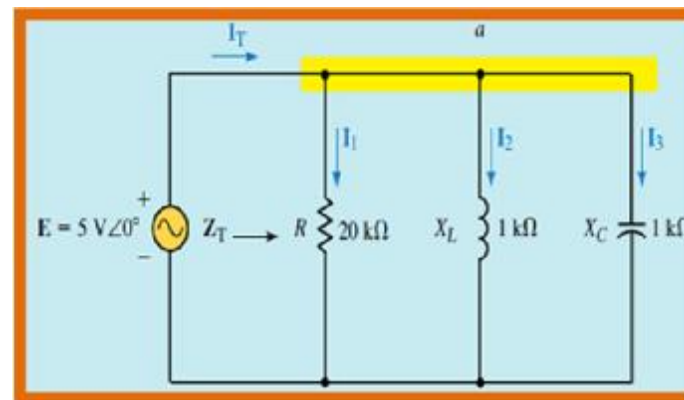
**Ex. 6:** Refer to the circuit;

(a) Find the total impedance,  $Z_T$ .

(b) Determine the supply current,  $I_T$ .

(c) Calculate  $I_1$ ,  $I_2$ , using current divider rule.

(d) Verify Kirchhoff's current law at node a.



### Solution

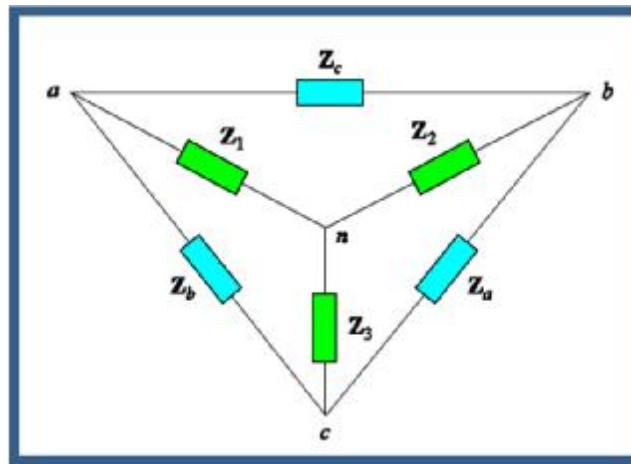
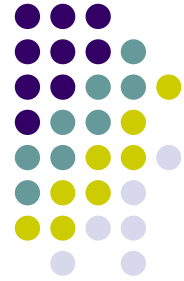
$$\begin{aligned} a. \quad \frac{1}{Z_T} &= \frac{1}{20} + \frac{1}{j1} + \frac{1}{-j1} = \frac{1}{20} & Z_T &= 20 \angle 0^\circ \text{ k}\Omega & Z_R &= 20 \angle 0^\circ \text{ k}\Omega \\ b. \quad I_T &= \frac{V}{Z_T} = \frac{5 \angle 0}{20 \times 10^3 \angle 0} = 250 \times 10^{-6} \angle 0^\circ \text{ A} = 250 \angle 0^\circ \text{ mA} & Z_L &= 1 \angle 90^\circ \text{ k}\Omega \\ & & Z_C &= 1 \angle -90^\circ \text{ k}\Omega \\ c. \quad I_1 &= \frac{20 \angle 0}{20 \angle 0} \times 250 \angle 0^\circ = 250 \angle 0^\circ \mu\text{A} \\ I_2 &= \frac{20 \angle 0}{1 \angle 90} \times 250 \angle 0^\circ = 5000 \angle -90^\circ \mu\text{A} \\ I_3 &= \frac{20 \angle 0}{1 \angle -90} \times 250 \angle 0^\circ = 5000 \angle 90^\circ \mu\text{A} \\ d. \quad I_T &= I_1 + I_2 + I_3 = 250 \angle 0 + 5000 \angle -90 + 5000 \angle 90 = 250 \angle 0^\circ \mu\text{A} \end{aligned}$$

$\therefore$  Kirchhoff's current law is verified.



# Delta to Star and Star to Delta Conversions

# Y- $\Delta$ and $\Delta$ -Y Equivalent Circuits



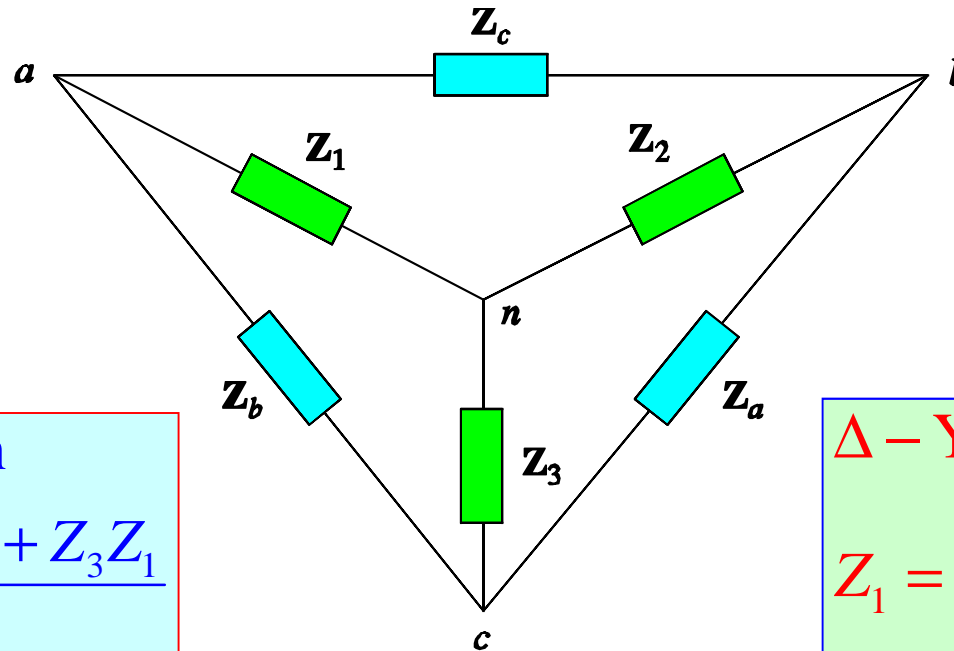
- Y- $\Delta$  and  $\Delta$ -Y type equivalent conversions will be most useful when considering Three Phase circuits.
- Impedances  $Z_1$ ,  $Z_2$  and  $Z_3$  are Y connected.
- Impedances  $Z_a$ ,  $Z_b$  and  $Z_c$  are  $\Delta$  connected.
- Y and  $\Delta$  forms can be equivalently converted from one form to the other.
- Y- $\Delta$  and  $\Delta$ -Y conversions are valid for impedances as well as resistive circuits.
- Y- $\Delta$  and  $\Delta$ -Y type equivalent conversions will be most useful when considering Three Phase circuits.



# Y- $\Delta$ and $\Delta$ -Y Equivalent Circuits



Y- $\Delta$  and  $\Delta$ -Y type equivalent conversions will be useful when considering Three Phase circuits.



## Y - $\Delta$ Conversion

$$Z_a = \frac{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}{Z_1}$$

$$Z_b = \frac{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}{Z_2}$$

$$Z_c = \frac{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}{Z_3}$$

## $\Delta$ - Y Conversion

$$Z_1 = \frac{Z_b Z_c}{Z_a + Z_b + Z_c}$$

$$Z_2 = \frac{Z_a Z_c}{Z_a + Z_b + Z_c}$$

$$Z_3 = \frac{Z_a Z_b}{Z_a + Z_b + Z_c}$$

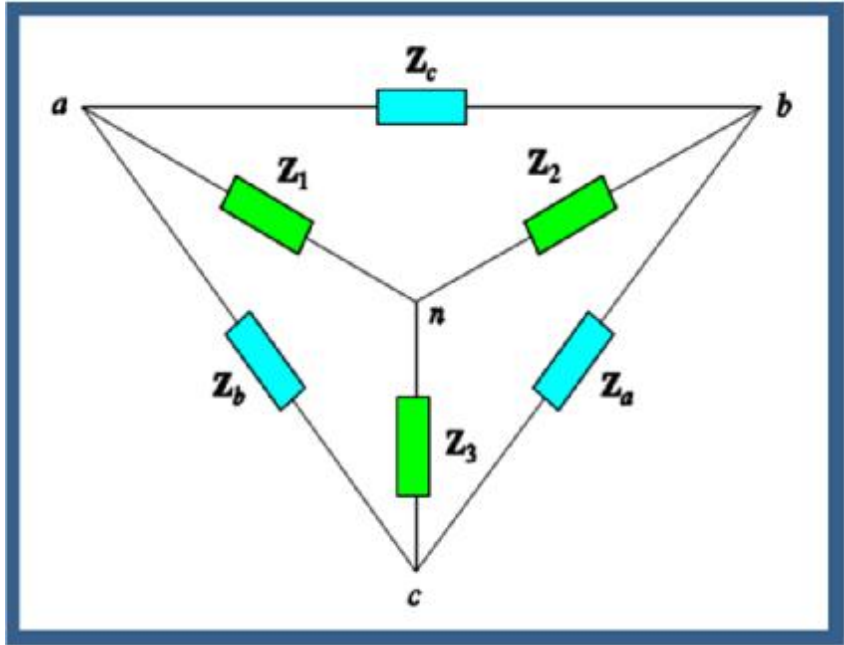
# Balanced Y- $\Delta$ and $\Delta$ -Y Equivalent Circuits



- Ø A delta ( $\Delta$ ) or Y (wye) circuit is balanced if it has equal impedances in all three branches.
- Ø Y- $\Delta$  and  $\Delta$ -Y conversions is very simple for balanced circuits.

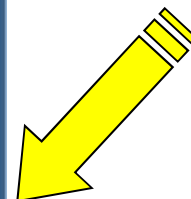
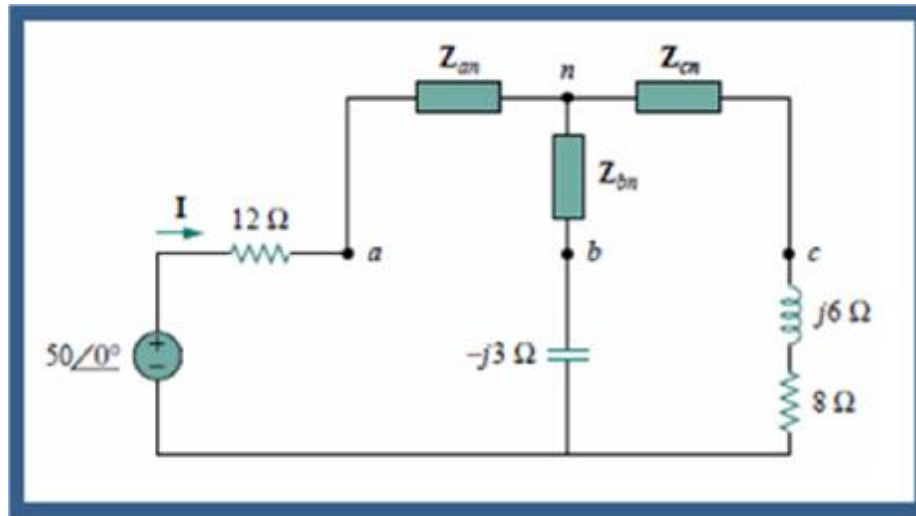
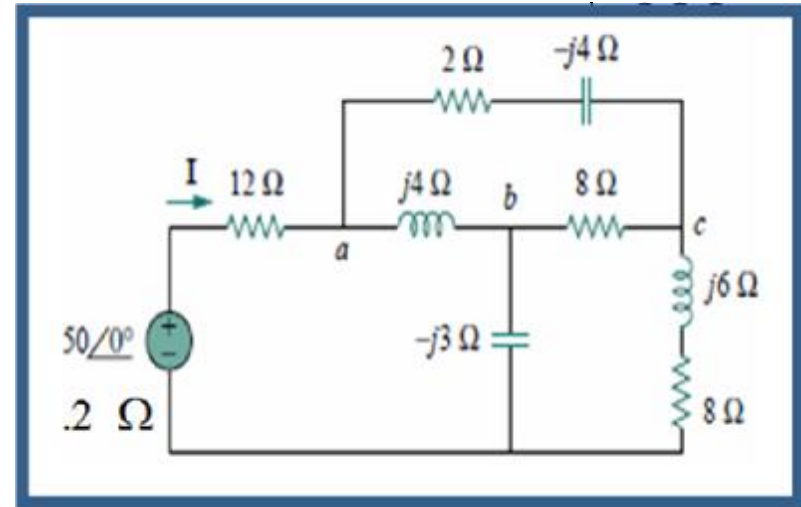
Balanced Impedance Conversions:  
 $Z_Y = Z_1 = Z_2 = Z_3$   
 $Z_\Delta = Z_a = Z_b = Z_c$

$$Z_\Delta = 3Z_Y \quad Z_Y = \frac{1}{3} Z_\Delta$$



**Ex. 7: Find the current I in the circuit.**

**Solution**

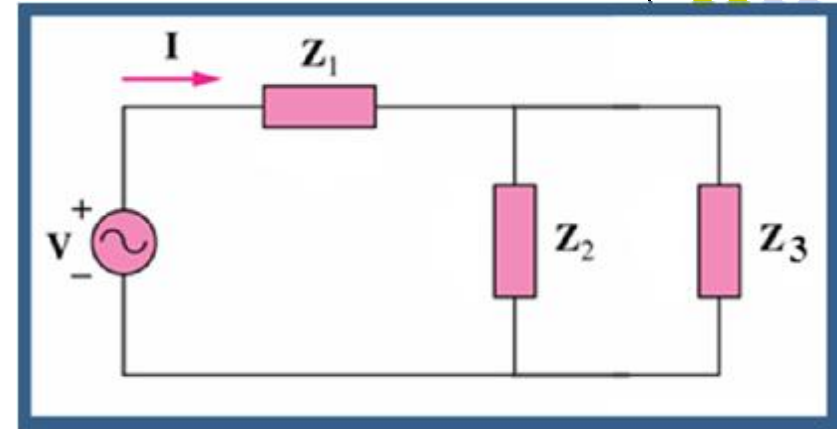
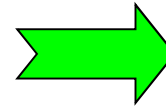
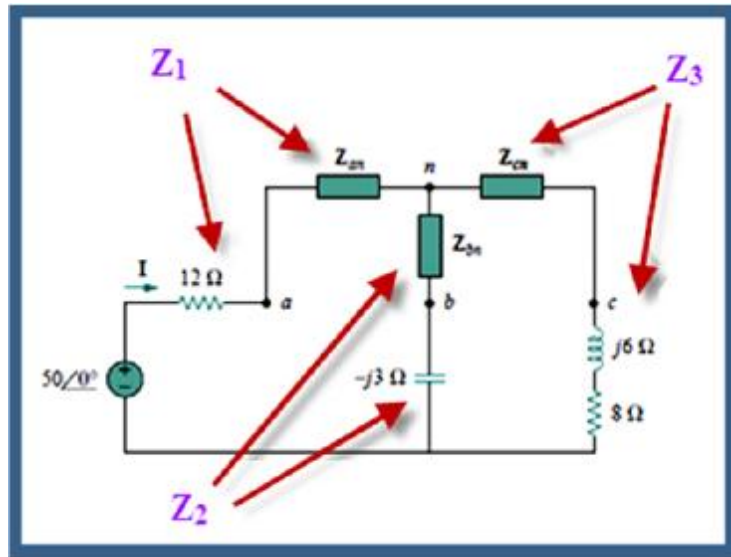


$$Z_{an} = \frac{j4(2 - j4)}{j4 + 2 - j4 + 8} = \frac{16 + j8}{10} = 1.6 + j0.8 \Omega$$

$$Z_{bn} = \frac{j4(8)}{10} = j3.2 \Omega$$

$$Z_{cn} = \frac{8(2 - j4)}{10} = 1.6 - j3.2 \Omega$$

**Ex. 7:** Find the current  $I$  in the circuit.



$$Z_1 = Z_{an} + 12 = 13.6 + j0.8$$

$$Z_2 = Z_{bn} - j3 = j0.2$$

$$Z_3 = Z_{cn} + j6 + 8 = 9.6 + j2.8$$

$$Z_T = Z_1 + \frac{Z_2 Z_3}{Z_2 + Z_3} = 13.6 + j0.8 + \frac{j0.2(9.6 + j2.8)}{9.6 + j3} = 13.6 + j1 \Omega$$

The source current  $I = \frac{V}{Z_T} = \frac{50 \angle 0}{13.6 + j1} = 3.666 \angle -4.2^\circ \Omega$



# AC Bridges

# AC Bridges

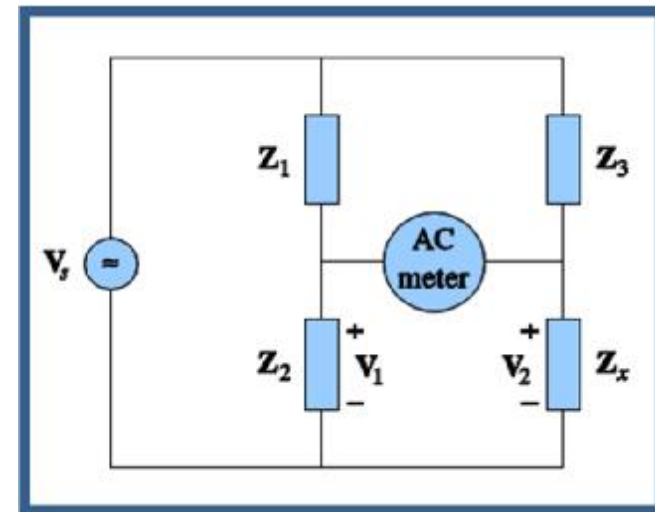


- ∅ The AC bridge is **Balanced** when no current flows through the meter. AC bridges are **used in measuring inductance and capacitance** values.

**At balance case:  $I_{\text{meter}} = 0$**

$$V_1 = V_2$$
$$V_1 = \frac{Z_2}{Z_1 + Z_2} V_s = V_2 = \frac{Z_x}{Z_3 + Z_x} V_s$$
$$\frac{Z_2}{Z_1 + Z_2} = \frac{Z_x}{Z_3 + Z_x} \quad \Rightarrow \quad Z_2 Z_3 = Z_1 Z_x$$

$$Z_x = \frac{Z_3}{Z_1} Z_2$$



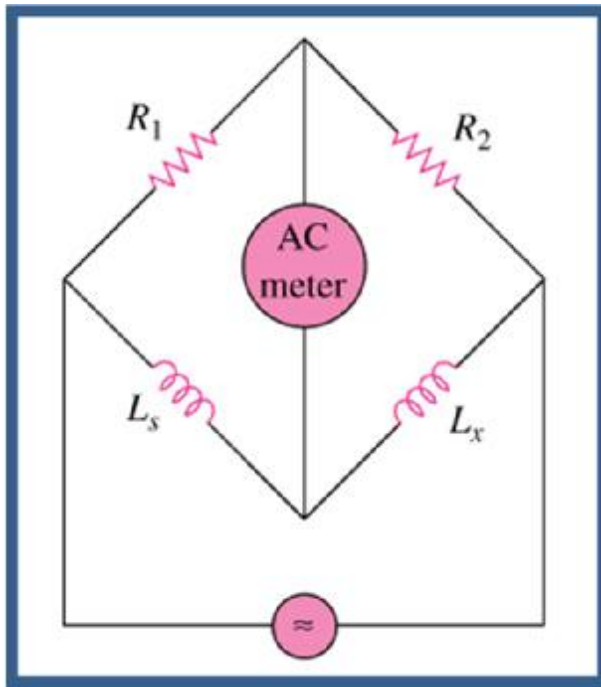
**AC bridge circuit**

**$Z_x$  Unknown value necessary for balancing the bridge**

# AC Bridges

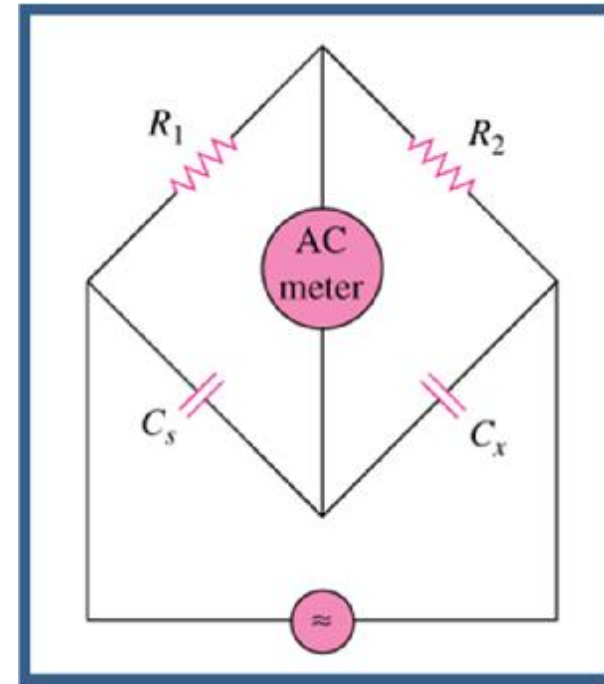


Unknown capacitance and inductances  $C_x$  and  $L_x$  are measured in terms of the known standard values  $C_s$  and  $L_s$



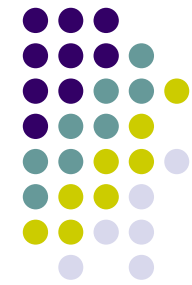
AC Bridge for measuring L

$$L_x = \frac{R_2}{R_1} L_s$$



AC Bridge for measuring C.

$$C_x = \frac{R_1}{R_2} C_s$$



**Ex.8:** Determine the series components that make up  $Z_x$  for balancing the bridge. Assume  $Z_1 = 4.8 \text{ k}\Omega$  resistor,  $Z_2$  is  $10 \text{ }\Omega$  in series with  $0.25 \text{ }\mu\text{H}$ ,  $Z_3 = 12 \text{ k}\Omega$  and  $f = 6 \text{ Mhz}$ .

### Solution

$$Z_x = (Z_3 / Z_1) Z_2 \quad Z_1 = 4.8 \text{ k}\Omega$$

$$Z_3 = 12 \text{ k}\Omega$$

$$Z_2 = 10 + j\omega L$$

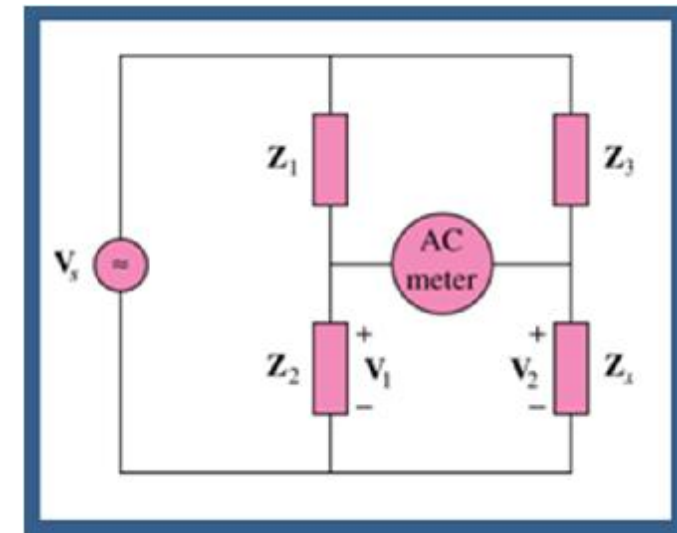
$$= 10 + j(2\pi)(6 \times 10^6)(0.25 \times 10^{-6})$$

$$= 10 + j9.425 \text{ k}\Omega$$

$$Z_x = \frac{12\text{k}}{4.8\text{k}}(10 + j9.425) = 25 + j23.5625 \text{ }\Omega$$

$$R_x = 25, \quad X_x = 23.5625 = \omega L_x$$

$$L_x = \frac{X_x}{2\pi f} = \frac{23.5625}{2\pi(6 \times 10^6)} = 0.625 \text{ }\mu\text{H}$$



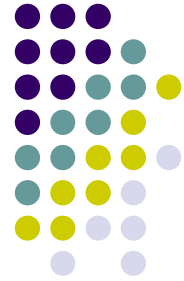
The series components are  $25 \text{ }\Omega$  with a  $0.625 \text{ }\mu\text{H}$





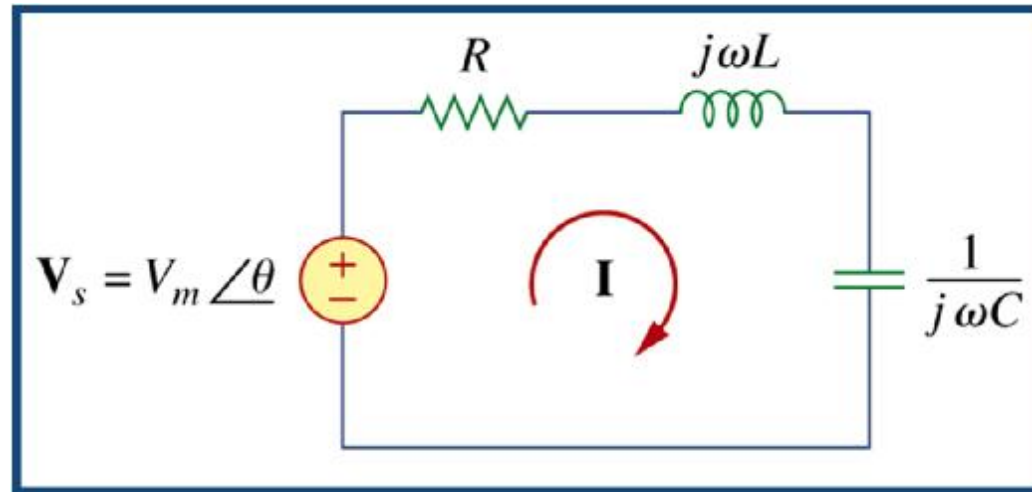
# Resonance in AC Circuits

# Resonance in AC Circuits



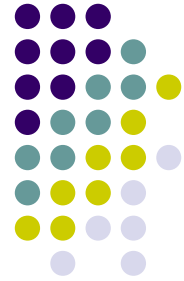
- I Resonance is a condition in an **RLC circuit** in which the **capacitive and reactive reactance** are **equal in magnitude**, the result is a purely **resistive impedance**.
- I Resonance circuits are useful for constructing **filters** and used in many application as **signal processing** and **communications systems** .

# Series Resonant Circuit



- At resonance, the impedance consists only **resistive component  $R$** .
- The value of current will be **maximum** since the total impedance is **minimum**.
- The voltage and current are **in phase**.
- Maximum power occurs at resonance since the power factor is **unity**.

# Series Resonant Circuit



Total impedance of series RLC Circuit is

$$Z_{\text{Total}} = R + jX_L - jX_C$$

$$Z_{\text{Total}} = R + j(X_L - X_C)$$

At Resonance:

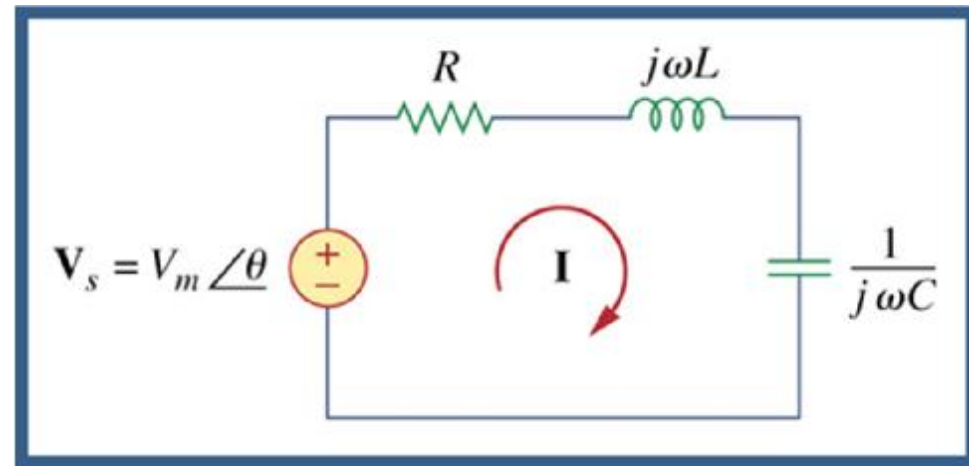
$$X_L = X_C$$

The impedance now reduce to

$$Z_{\text{Total}} = R$$

The current at resonance

$$I_m = \frac{V_s}{Z_{\text{Total}}} = \frac{V_m}{R}$$



The highest power is at  $\omega_o$

$$P(\omega_o) = I^2 R = \frac{V^2}{R}$$

# Series Resonance



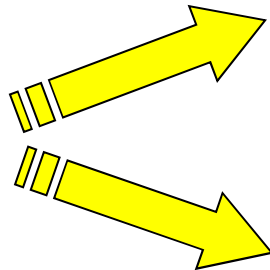
## Resonance Frequency

Resonance frequency is the frequency where the condition of resonance occur.

$$\text{Im}(z) = 0 \quad \longrightarrow \quad X_L = X_C$$

$$wL = \frac{1}{wC} \quad \longrightarrow \quad wL - \frac{1}{wC} = 0 \quad \longrightarrow \quad w_o L = \frac{1}{w_o C}$$

Resonance Frequency


$$\omega_o = \frac{1}{\sqrt{LC}} \quad \text{rad/sec}$$
$$f_o = \frac{1}{2\pi\sqrt{LC}} \quad \text{Hz}$$

# Series Resonance



## Half-power Frequency $\omega_1$ & $\omega_2$

Half-power frequencies is the frequency when the magnitude of the output voltage or current is decrease by the factor of  $1 / \sqrt{2}$  from its maximum value. Also known as **cutoff frequencies**.

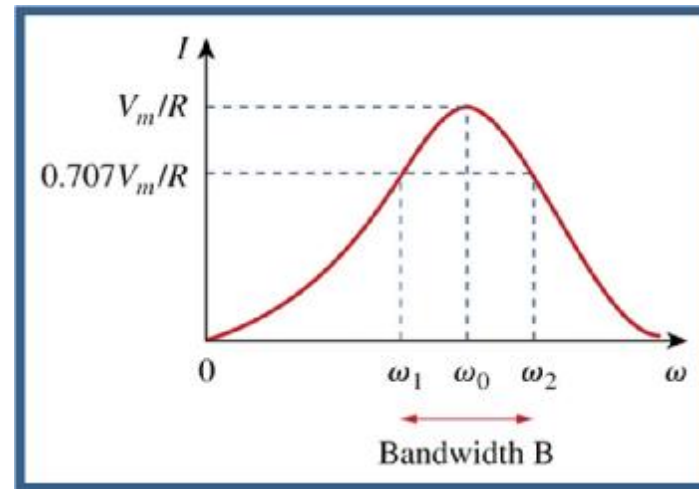
$$I = \frac{I}{\sqrt{2}} = 0.707I$$

- The half-power frequencies  $\omega_1$  and  $\omega_2$  can be obtained by setting,

$$|Z(\omega_1)| = |Z(\omega_2)| = \sqrt{R^2 + (\omega L - 1/\omega C)^2} = \sqrt{2}R$$

$$P(\omega_1) = P(\omega_2) = \frac{\left(\frac{V_m}{\sqrt{2}}\right)^2}{2R}$$

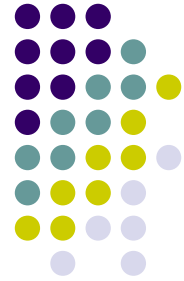
$$\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$



Response curve

$$\omega_2 = +\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

# Series Resonance



## Maximum Power Dissipated

The maximum current at resonance where:

$$I = \frac{V_m}{R}$$

Thus maximum power dissipated is:

$$P = I^2 R \quad \longrightarrow \quad \text{at } \omega = \omega_0$$

The average power dissipated by the RLC circuit is:

$$P = \frac{1}{2} I^2 R \quad P = \frac{1}{2} \frac{V_m^2}{R} \quad \longrightarrow \quad \text{at } \omega = \omega_1, \omega_2,$$

# Series Resonance



## Bandwidth of resonant circuit, $b$

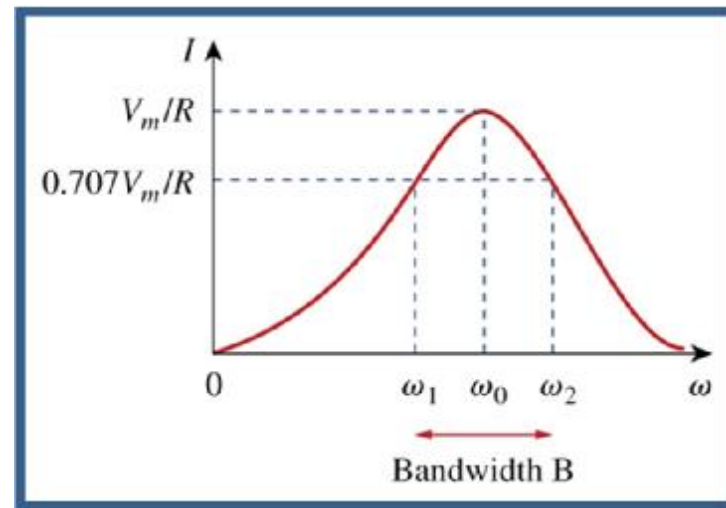
Bandwidth,  $b$  is define as the difference between the two half power frequencies. The width of the response curve is determine by the bandwidth.

$$B = \omega_2 - \omega_1$$

$$\beta = \frac{R}{L}$$

Resonance frequency can be obtained from the half-power frequencies.

$$\omega_o = \sqrt{\omega_1 \omega_2}$$



Response curve



# Quality Factor (Q-Factor)



Quality factor is used to measure the “sharpness” of response curve

$$Q = 2\pi \frac{\text{Peak Energy Stored}}{\text{Energy Dissipated in one Period at Resonance}}$$

Quality factor is the ratio of resonance frequency to the bandwidth

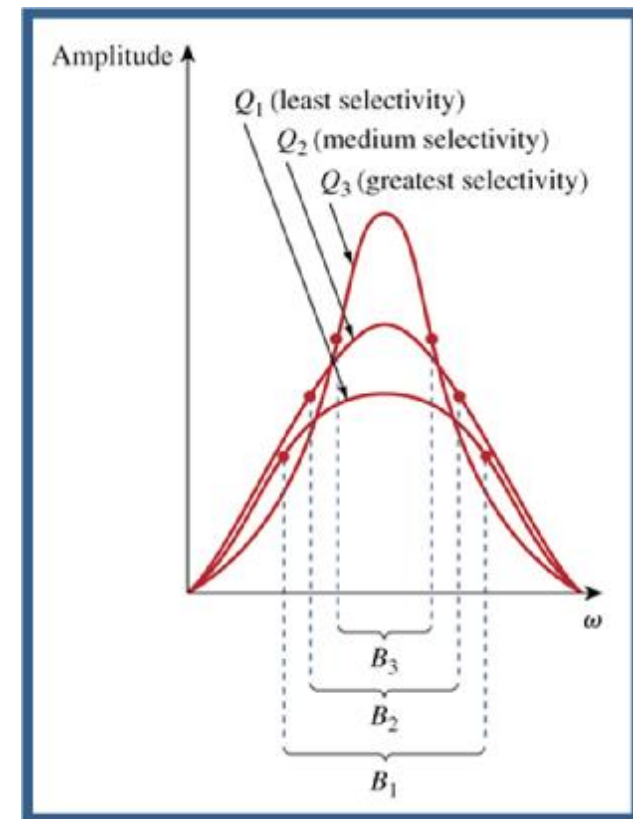
$$Q = \frac{\omega_o}{\beta}$$

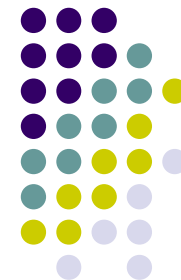
$$\beta = \frac{R}{L}$$

$$Q = \frac{\omega_o L}{R}$$

Selectivity defines how well a resonant circuit responds to certain frequencies.

- Higher value of  $Q$ , smaller the bandwidth. (Higher the selectivity)
- Lower value of  $Q$ , larger the bandwidth. (Lower the selectivity)





## High-Q

It is to be a **high-Q circuit** when its **quality factor** is equal or greater than 10.

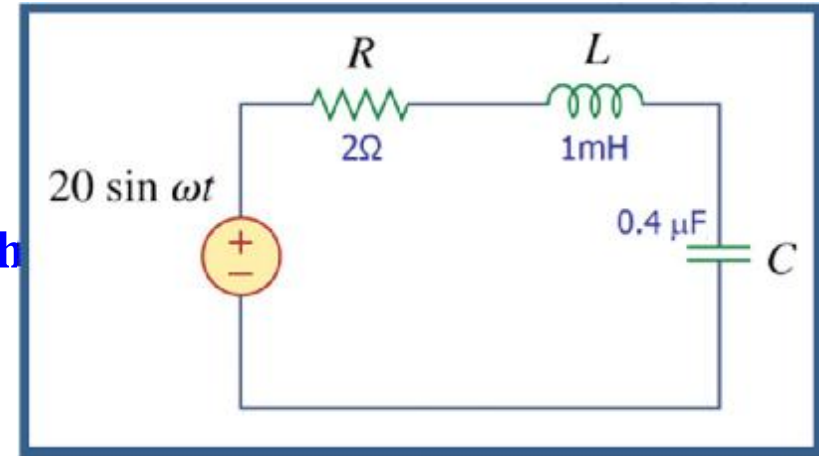
For a **high-Q circuit** ( $Q \geq 10$ ), the **half-power frequencies** are, for all practical purposes, symmetrical around the resonant frequency and can be approximated as

$$\omega_1 \cong \omega_o - \frac{\beta}{2}$$

$$\omega_2 \cong \omega_o + \frac{\beta}{2}$$

**Ex.9:** For the circuit shown:

- (a) Find the resonant frequency and the half power frequencies
- (b) Calculate the quality factor and bandwidth
- (c) Determine the amplitude of the current at  $\omega_0$ ,  $\omega_1$  and  $\omega_2$



**Solution**

$$(a) \quad \omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{10^{-3} \times 0.4 \times 10^{-6}}} = 50 \text{ krad/s}$$

$$\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}} = -\frac{2}{2 \times 10^{-3}} + \sqrt{(10^3)^2 + (50 \times 10^3)^2}$$

$$\omega_1 = -1 + \sqrt{1 + 2500} \text{ krad/s} = 49 \text{ krad/s} \quad \omega_2 = 1 + \sqrt{1 + 2500} \text{ krad/s} = 51 \text{ krad/s}$$

$$(b) \quad Q = \frac{\omega_0}{B} = \frac{50}{2} = 25 \quad B = \omega_2 - \omega_1 = 2 \text{ krad/s}$$

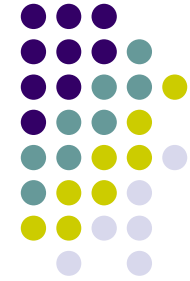
$$(c) \quad \underline{\text{At } \omega = \omega_0} \quad I = \frac{V_m}{R} = \frac{20}{2} = 10 \text{ A}$$

$$\underline{\text{At } \omega = \omega_1, \omega_2} \quad I = \frac{V_m}{\sqrt{2}R} = \frac{10}{\sqrt{2}} = 7.071 \text{ A}$$

Amplitude

$$i(t) = I_m \sin \omega t$$

# Parallel Resonant Circuit



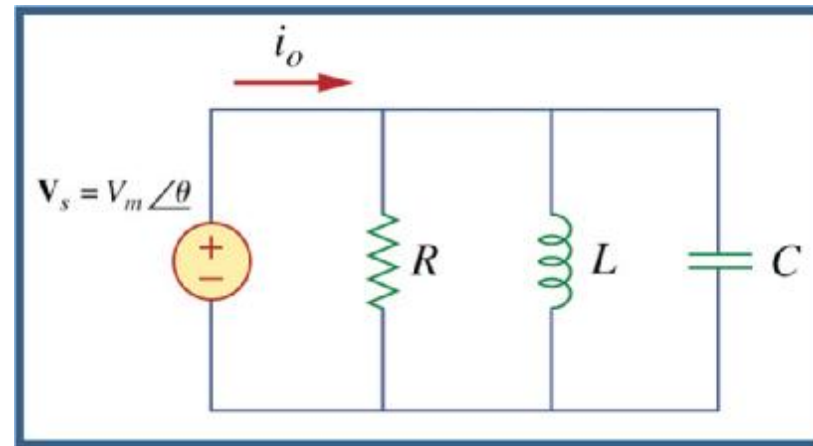
The total admittance:

$$Y_{\text{Total}} = Y_1 + Y_2 + Y_3$$

$$Y_{\text{Total}} = \frac{1}{R} + \frac{1}{j\omega L} + \frac{1}{-j/\omega C}$$

$$Y_{\text{Total}} = \frac{1}{R} + \frac{-j}{\omega L} + j\omega C$$

$$Y_{\text{Total}} = \frac{1}{R} + j(\omega C - 1/\omega L)$$



Resonance occur when:

$$\omega C = \frac{1}{\omega L}$$

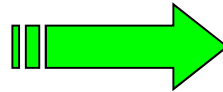
- At resonance, the impedance consists only conductance  $G$ .
- The value of current will be minimum since the total admittance is minimum.
- The voltage and current are in phase.

# Parameters in Parallel Circuit



Parallel resonant circuit has same parameters as the series resonant circuit.

Resonance frequency



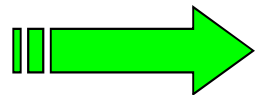
$$\omega_o = \frac{1}{\sqrt{LC}} \text{ rad/s}$$

Half-power frequencies

$$\omega_1 = \frac{-1}{2RC} + \sqrt{\left(\frac{1}{2RC}\right)^2 + \left(\frac{1}{LC}\right)} \text{ rad/s}$$

$$\omega_2 = \frac{1}{2RC} + \sqrt{\left(\frac{1}{2RC}\right)^2 + \left(\frac{1}{LC}\right)} \text{ rad/s}$$

Bandwidth

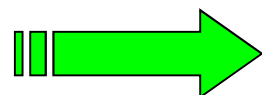


$$\beta = \omega_2 - \omega_1 = \frac{1}{RC}$$

For a high-Q circuit ( $Q \geq 10$ )

$$\omega_1 \cong \omega_o - \frac{\beta}{2}$$

Q - Factor



$$Q = \frac{\omega_o}{\beta} = \omega_o RC = \frac{R}{\omega_o L}$$

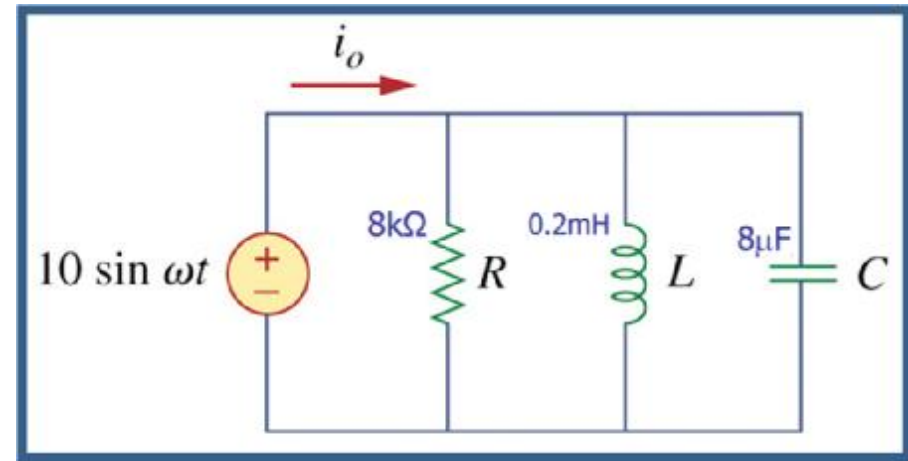
$$\omega_2 \cong \omega_o + \frac{\beta}{2}$$

**Ex.10:** For the circuit shown:

(a) Calculate  $\omega_0$ ,  $Q$  and  $B$

(b) Find  $\omega_1$  and  $\omega_2$

(c) Determine the power dissipated at  $\omega_0$ ,  $\omega_1$  and  $\omega_2$



**Solution**

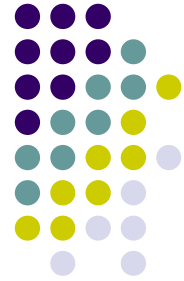
$$(a) \quad \omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{0.2 \times 10^{-3} \times 8 \times 10^{-6}}} = \frac{10^5}{4} = 25 \text{ krad/s}$$

$$Q = \frac{R}{\omega_0 L} = \frac{8 \times 10^3}{25 \times 10^3 \times 0.2 \times 10^{-3}} = 1600$$

$$B = \frac{\omega_0}{Q} = 15.625 \text{ rad/s}$$

$$(b) \quad \omega_1 = \omega_0 - \frac{B}{2} = 25,000 - 7.812 = 24,992 \text{ rad/s}$$

$$\omega_2 = \omega_0 + \frac{B}{2} = 25,000 + 7.8125 = 25,008 \text{ rad/s}$$



At  $\omega = \omega_0$ ,  $Y = 1/R$  or  $Z = R = 8 \text{ k}\Omega$ .

$$I_o = \frac{V}{Z} = \frac{10 \angle -90^\circ}{8000} = 1.25 \angle -90^\circ \text{ mA}$$

at  $\omega = \omega_0$

$$P = I_o^2 R = (1.25 \times 10^{-3})^2 (8000) = 0.0125 \text{ W} = \boxed{12.5 \times 10^{-3} \text{ W}}$$

at  $\omega = \omega_1, \omega_2$ ,

$$P(\omega_o) = \frac{1}{2} \frac{V_m^2}{R}$$

$$P = \frac{V^2}{2R} = \frac{10^2}{2 \times 8000} = \boxed{6.25 \times 10^{-3} \text{ W}}$$